

## LM108/LM208/LM308 Operational Amplifiers

Check for Samples: [LM108-N](#), [LM208-N](#), [LM308-N](#)

### FEATURES

- Maximum input bias current of 3.0 nA over temperature
- Offset current less than 400 pA over temperature
- Supply current of only 300  $\mu$ A, even in saturation
- Guaranteed drift characteristics

### DESCRIPTION

The LM108 series are precision operational amplifiers having specifications a factor of ten better than FET amplifiers over a  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$  temperature range.

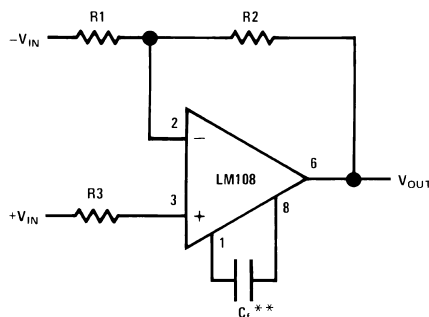
The devices operate with supply voltages from  $\pm 2\text{V}$  to  $\pm 20\text{V}$  and have sufficient supply rejection to use unregulated supplies. Although the circuit is interchangeable with and uses the same compensation as the LM101A, an alternate compensation scheme can be used to make it particularly insensitive to power supply noise and to make supply bypass capacitors unnecessary.

The low current error of the LM108 series makes possible many designs that are not practical with conventional amplifiers. In fact, it operates from  $10\text{ M}\Omega$  source resistances, introducing less error than devices like the 709 with  $10\text{ k}\Omega$  sources. Integrators with drifts less than  $500\text{ }\mu\text{V}/\text{sec}$  and analog time delays in excess of one hour can be made using capacitors no larger than  $1\text{ }\mu\text{F}$ .

The LM108 is guaranteed from  $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ , the LM208 from  $-25^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ , and the LM308 from  $0^{\circ}\text{C}$  to  $+70^{\circ}\text{C}$ .

### Compensation Circuits

Improves rejection of power supply noise by a factor of ten.



$$C_f \geq \frac{R_1 C_O}{R_1 + R_2}$$

$$C_O = 30\text{ pF}$$

\*\*Bandwidth and slew rate are proportional to  $1/C_f$

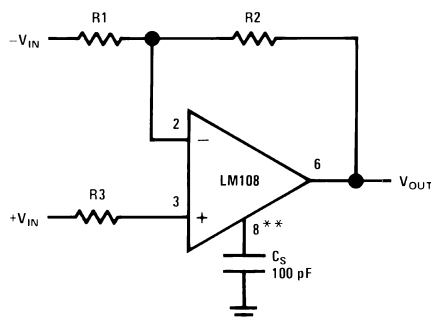
**Figure 1. Standard Compensation Circuit**



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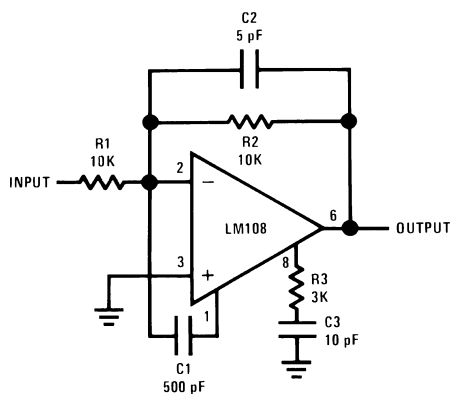
Improves rejection of power supply noise by a factor of ten.



\*\*Bandwidth and slew rate are proportional to  $1/C_s$

1. Improves rejection of power supply noise by a factor of ten.

**Figure 2. Alternate Frequency Compensation<sup>(1)</sup>**



**Figure 3. Feedforward Compensation**



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

**Absolute Maximum Ratings** <sup>(1) (2)</sup>

	LM108/LM208	LM308
Supply Voltage	±20V	±18V
Power Dissipation <sup>(3)</sup>	500 mW	500 mW
Differential Input Current <sup>(4)</sup>	±10 mA	±10 mA
Input Voltage <sup>(5)</sup>	±15V	±15V
Output Short-Circuit Duration	Continuous	Continuous
Operating Temperature Range (LM108)	-55°C to +125°C	0°C to +70°C
(LM208)	-25°C to + 85°C	
Storage Temperature Range	-65°C to +150°C	-65°C to +150°C
Lead Temperature (Soldering, 10 sec)		
DIP	260°C	260°C
H Package Lead Temp, (Soldering 10 seconds)	300°C	300°C
Soldering Information, Dual-In-Line Package, Soldering (10 seconds)	260°C	
Small Outline Package		
Vapor Phase (60 seconds)	215°C	
Infrared (15 seconds)	220°C	
See AN-450 "Surface Mounting Methods and Their Effect on Product Reliability" for other methods of soldering surface mount devices.		
ESD Tolerance <sup>(6)</sup>	2000V	

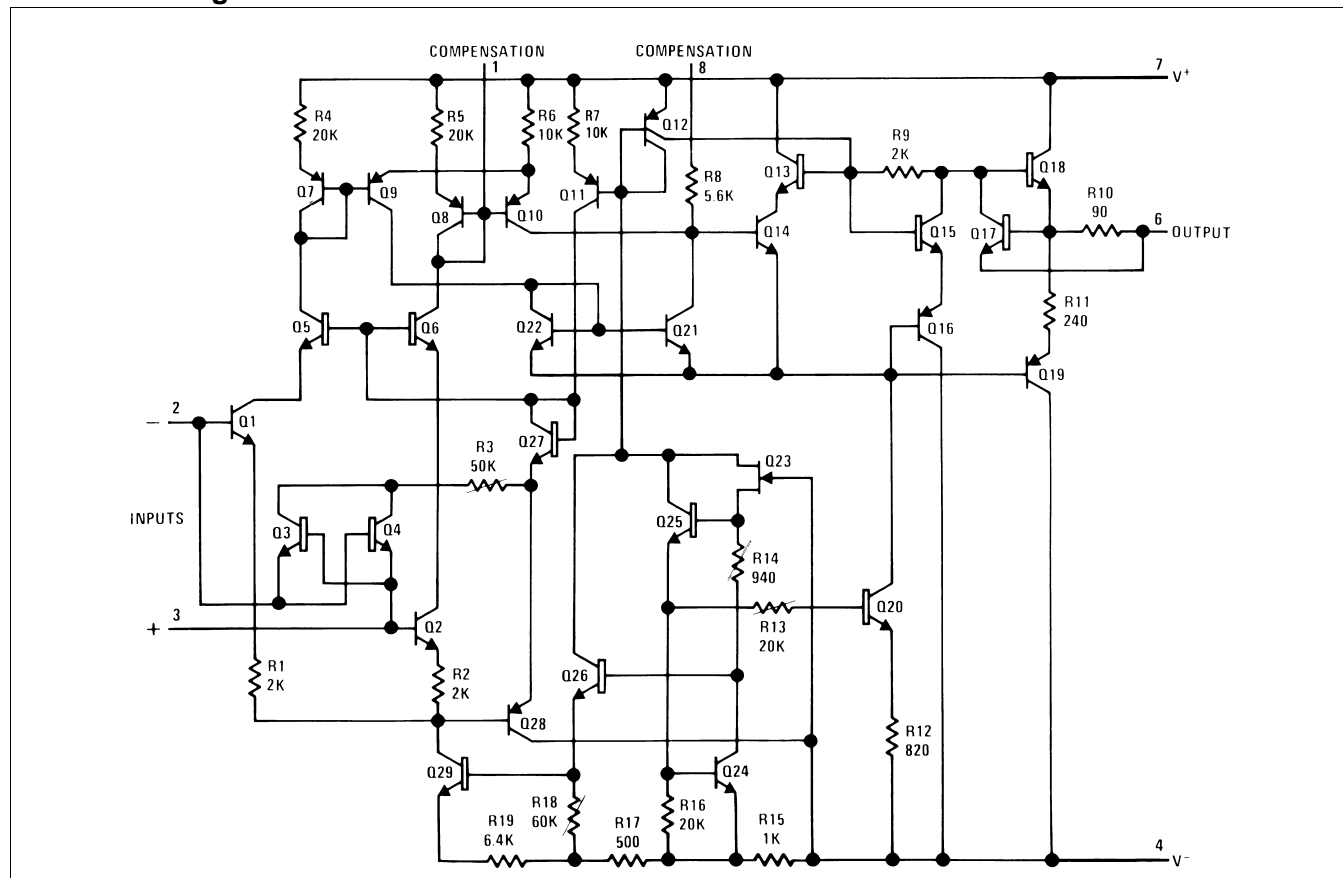
- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits.
- (2) Refer to RETS108X for LM108 military specifications and RETs 108AX for LM108A military specifications.
- (3) The maximum junction temperature of the LM108 is 150°C, for the LM208, 100°C and for the LM308, 85°C. For operating at elevated temperatures, devices in the H08 package must be derated based on a thermal resistance of 160°C/W, junction to ambient, or 20°C/W, junction to case. The thermal resistance of the dual-in-line package is 100°C/W, junction to ambient.
- (4) The inputs are shunted with back-to-back diodes for overvoltage protection. Therefore, excessive current will flow if a differential input voltage in excess of 1V is applied between the inputs unless some limiting resistance is used.
- (5) For supply voltages less than ±15V, the absolute maximum input voltage is equal to the supply voltage.
- (6) Human body model, 1.5 kΩ in series with 100 pF.

**Electrical Characteristics** <sup>(1)</sup>

Parameter	Condition	LM108/LM208			LM308			Units
		Min	Typ	Max	Min	Typ	Max	
Input Offset Voltage	$T_A = 25^\circ\text{C}$		0.7	2.0		2.0	7.5	mV
Input Offset Current	$T_A = 25^\circ\text{C}$		0.05	0.2		0.2	1	nA
Input Bias Current	$T_A = 25^\circ\text{C}$		0.8	2.0		1.5	7	nA
Input Resistance	$T_A = 25^\circ\text{C}$	30	70		10	40		M $\Omega$
Supply Current	$T_A = 25^\circ\text{C}$		0.3	0.6		0.3	0.8	mA
Large Signal Voltage	$T_A = 25^\circ\text{C}, V_S = \pm 15\text{V}$	50	300		25	300		V/mV
Gain	$V_{\text{OUT}} = \pm 10\text{V}, R_L \geq 10\text{ k}\Omega$							
Input Offset Voltage				3.0			10	mV
Average Temperature								
Coefficient of Input			3.0	15		6.0	30	$\mu\text{V}/^\circ\text{C}$
Offset Voltage								
Input Offset Current				0.4			1.5	nA
Average Temperature								
Coefficient of Input			0.5	2.5		2.0	10	$\text{pA}/^\circ\text{C}$
Offset Current								
Input Bias Current				3.0			10	nA
Supply Current	$T_A = +125^\circ\text{C}$		0.15	0.4				mA
Large Signal Voltage	$V_S = \pm 15\text{V}, V_{\text{OUT}} = \pm 10\text{V}$	25			15			V/mV
Gain	$R_L \geq 10\text{ k}\Omega$							
Output Voltage Swing	$V_S = \pm 15\text{V}, R_L = 10\text{ k}\Omega$	$\pm 13$	$\pm 14$		$\pm 13$	$\pm 14$		V
Input Voltage Range	$V_S = \pm 15\text{V}$	$\pm 13.5$			$\pm 14$			V
Common Mode		85	100		80	100		dB
Rejection Ratio								
Supply Voltage		80	96		80	96		dB
Rejection Ratio								

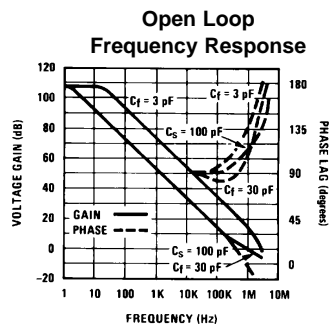
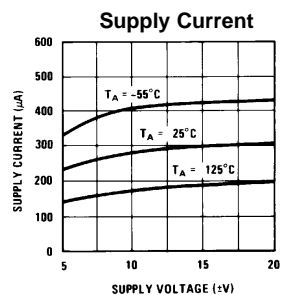
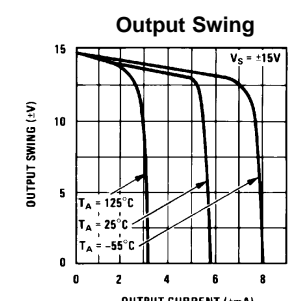
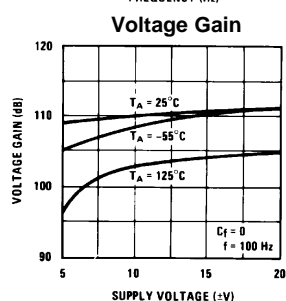
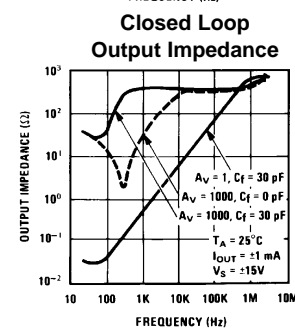
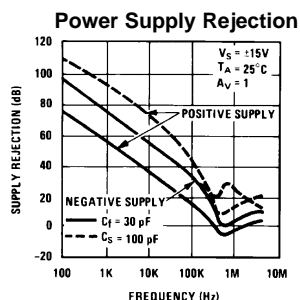
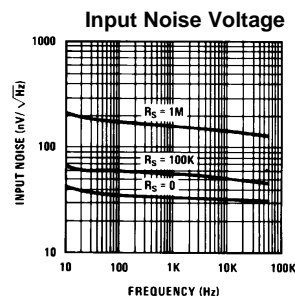
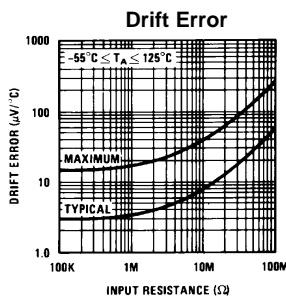
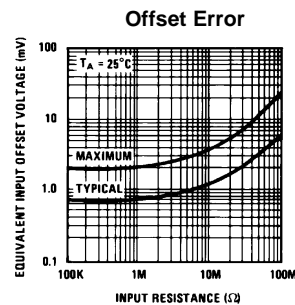
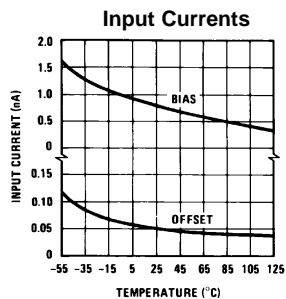
(1) These specifications apply for  $\pm 5\text{V} \leq V_S \leq \pm 20\text{V}$  and  $-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$ , unless otherwise specified. With the LM208, however, all temperature specifications are limited to  $-25^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$ , and for the LM308 they are limited to  $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ .

Schematic Diagram



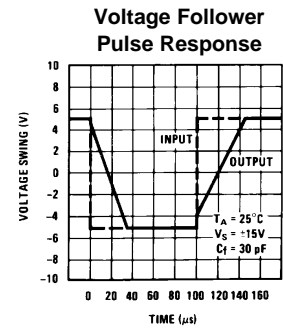
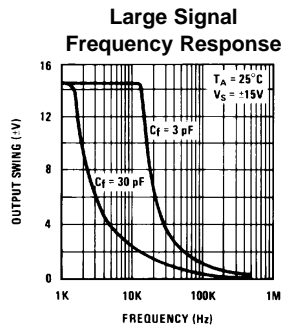
## Typical Performance Characteristics

LM108/LM208



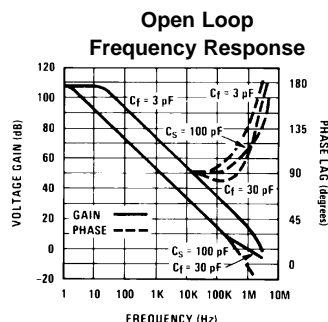
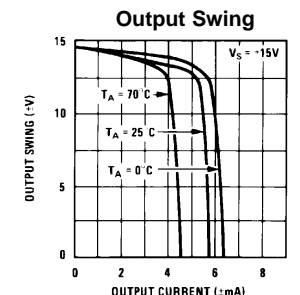
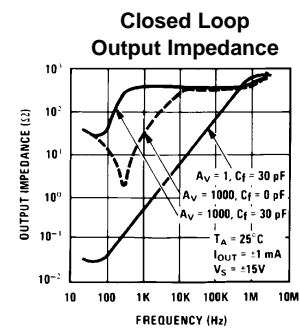
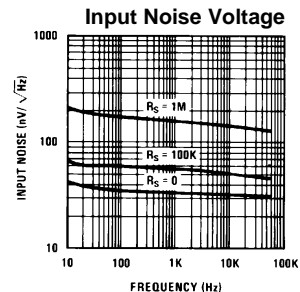
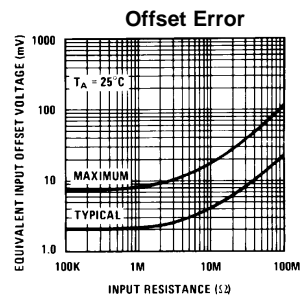
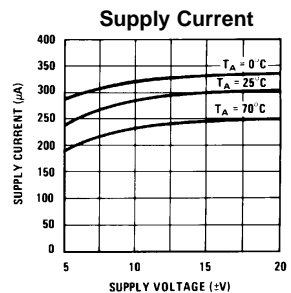
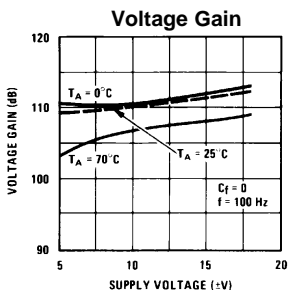
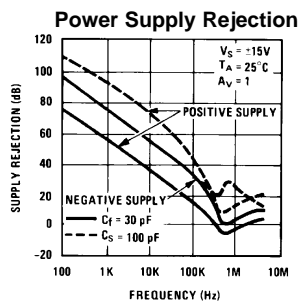
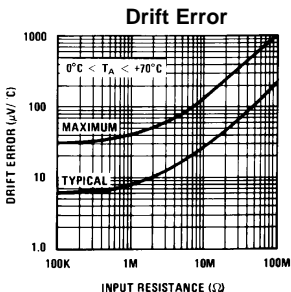
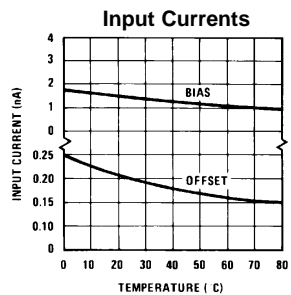
**Typical Performance Characteristics (continued)**

LM108/LM208



## Typical Performance Characteristics

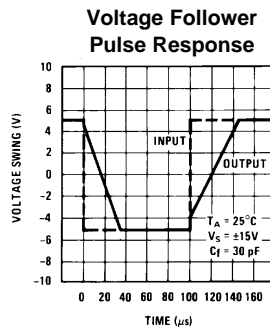
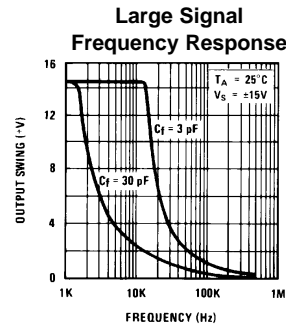
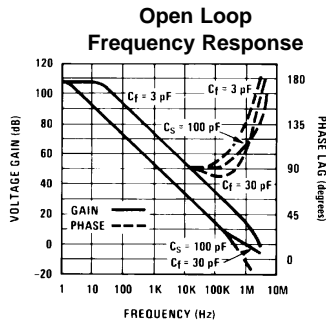
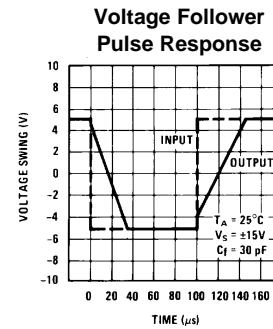
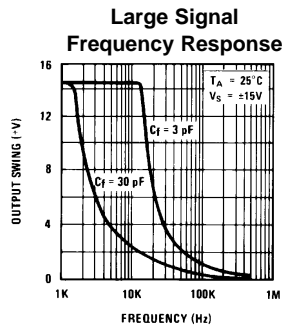
### LM308



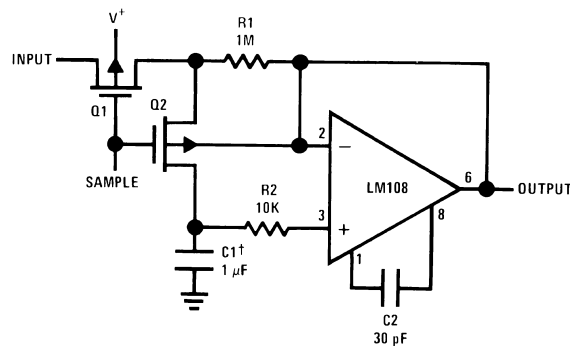


Typical Performance Characteristics (continued)

LM308

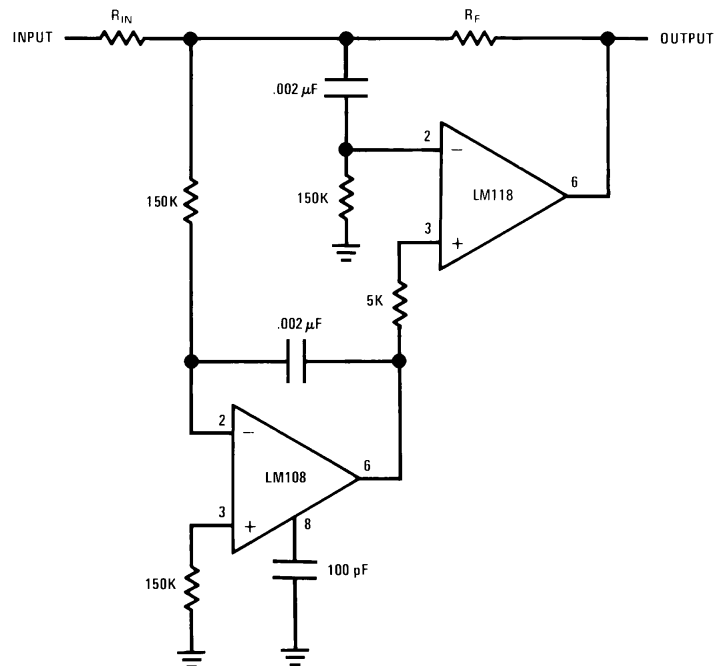


Typical Applications

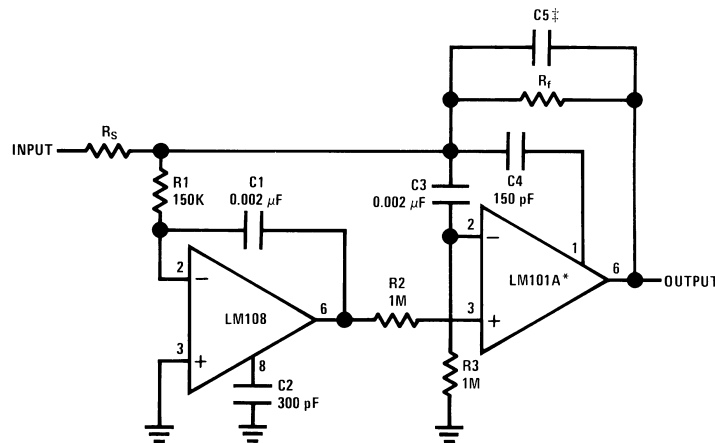


†Teflon polyethylene or polycarbonate dielectric capacitor  
 Worst case drift less than 2.5 mV/sec

Figure 4. Sample and Hold



**Figure 5. High Speed Amplifier with Low Drift and Low Input Current**



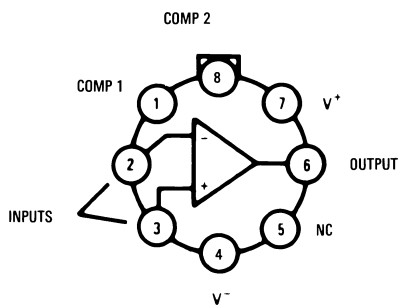
$$C5 = \frac{6 \times 10^{-8}}{R_f}$$

\*In addition to increasing speed, the LM101A raises high and low frequency gain, increases output drive capability and eliminates thermal feedback.

1. Power Bandwidth: 250 KHz, Small Signal Bandwidth: 3.5 MHz, Slew Rate: 10V/μS

**Figure 6. Fast Summing Amplifier<sup>(1)</sup>**

Connection Diagrams



- (1) Package is connected to Pin 4 ( $V^-$ )
- (2) Unused pin (no internal connection) to allow for input anti-leakage guard ring on printed circuit board layout.

Figure 7. Metal Can Package<sup>(1)(2)</sup>

Dual-In-Line Package

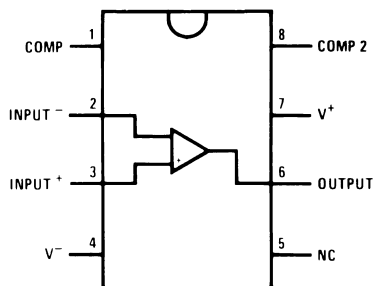


Figure 8. Top View (8-Pin)

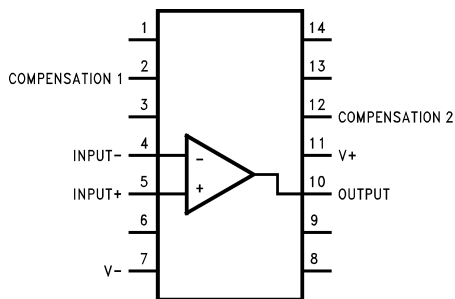
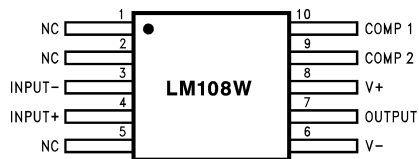


Figure 9. Top View (14-Pin)



†Also available per JM38510/10104

Figure 10. Top View (10-Pin)

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